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(54) MICROWAVE ABLATION ANTENNA SYSTEM WITH TAPERED SLOT BALUN

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(57) **ABSTRACT**

A balun includes a center conductor, a dielectric material, a tapered wall, a ring, and a prong. The center conductor extends a length of the balun. The dielectric material surrounds the center conductor along the length of the balun. The tapered wall forms a portion of a tube between a first wall and a second wall. The first wall is opposite the second wall. The tapered wall is formed of a conductive material. The portion of the tube forms a slot exposing the dielectric material. The ring connects to the second wall of the tapered wall and is formed of the conductive material. The ring forms a tube surrounding the center conductor and the dielectric material. The prong connects to the ring to extend toward the first wall and is formed of the conductive material. The prong extends over a portion of the dielectric material exposed by the slot.

20 Claims, 28 Drawing Sheets



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FIG. 9





FIG. 10























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FIG. 20



Antenna <u>100</u>	2200	
Balun <u>300</u> , <u>1100</u>		
Coaxial cable <u>102</u>		
Connector 2206		
Amplifier 2204		
Signal generator <u>2202</u>		



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MICROWAVE ABLATION ANTENNA SYSTEM WITH TAPERED SLOT BALUN

REFERENCE TO GOVERNMENT RIGHTS

This invention was made with government support under ECCS-1406090 awarded by the National Science Foundation. The government has certain rights in the invention.

BACKGROUND

Microwave ablation (MWA) is a form of thermal ablation used in interventional radiology to treat cancer. MWA uses electromagnetic waves in the microwave energy spectrum (300 megahertz to 300 gigahertz) to produce tissue-heating 15 effects, i.e., to heat tumors to cytotoxic temperatures. MWA is generally used for minimally invasive treatment and/or palliation of solid tumors in patients. MWA offers several advantages over other ablation technologies such as radiofrequency (RF) and cryoablation including higher tempera- 20 tures than RF, larger ablation zone volumes, shorter ablation times, and better ablation performance near arteries, which act as heat sinks. Selective delivery of energy to the prescribed tissue volume (i.e. the tumor and its margins) is achieved by means of interstitial placement of a microwave 25 antenna directly into the tumor. Current MWA technology may be employed either laparoscopically or percutaneously, and thus, is considered to be minimally invasive. However, the extent to which MWA is minimally invasive depends on a length and a diameter of the interstitial microwave 30 antenna.

SUMMARY

A balun is provided that includes, but is not limited to, a 35 an illustrative embodiment. center conductor, a dielectric material, a tapered wall, a ring, and a prong. The center conductor extends a length of the balun. The dielectric material surrounds the center conductor along the length of the balun. The tapered wall forms a portion of a tube between a first wall and a second wall. The 40 FIGS. 3 and 4 unrolled and placed on a flat surface in first wall is opposite the second wall. The tapered wall is formed of a conductive material. The portion of the tube forms a slot exposing the dielectric material. The ring connects to the second wall of the tapered wall and is formed of the conductive material. The ring forms a tube surround- 45 ing the center conductor and the dielectric material. The prong connects to the ring to extend toward the first wall and is formed of the conductive material. The prong extends over a portion of the dielectric material exposed by the slot.

An antenna system is provided. The antenna system 50 includes, but is not limited to, a coaxial cable, a first dipole arm, and a balun. The coaxial cable includes, but is not limited to, a center conductor extending a length of the coaxial cable, a dielectric material surrounding the center conductor along the length of the coaxial cable, and a 55 conductive shield surrounding the dielectric material along the length of the coaxial cable. The first dipole arm connects to the center conductor. The balun connects between an end of the coaxial cable and the first dipole arm. The balun includes, but is not limited to, a second center conductor, a 60 second dielectric material, a tapered wall, a ring, and a prong. The second center conductor extends a length of the balun and connects to and extends from the center conductor between the center conductor and the first dipole arm. The second dielectric material surrounds the second center con- 65 ductor along the length of the balun. The tapered wall forms a portion of a tube between a first wall and a second wall,

wherein the first wall is opposite the second wall. The first wall is connected to the conductive shield. The tapered wall is formed of a conductive material. The portion of the tube forms a slot exposing the second dielectric material. The ring connects to the second wall of the tapered wall and is formed of the conductive material. The ring forms a tube surrounding the second center conductor and the second dielectric material. The prong connects to the ring to extend toward the first wall and is formed of the conductive material. The prong extends over a portion of the second dielectric material exposed by the slot and forms a second dipole arm.

Other principal features of the disclosed subject matter will become apparent to those skilled in the art upon review of the following drawings, the detailed description, and the appended claims.

BRIEF DESCRIPTION OF THE DRAWINGS

Illustrative embodiments of the disclosed subject matter will hereafter be described referring to the accompanying drawings, wherein like numerals denote like elements.

FIG. 1 depicts a microwave ablation (MWA) antenna system in accordance with an illustrative embodiment.

FIG. 2 depicts an MWA antenna system with a prior balun design in accordance with an illustrative embodiment.

FIG. 3 depicts a top view of an MWA antenna system with a first balun in accordance with an illustrative embodiment.

FIG. 4 depicts a right side view of the MWA antenna system of FIG. 3 in accordance with an illustrative embodiment.

FIG. 5A depicts the top and right side views of the MWA antenna system of FIGS. 3 and 4 aligned in accordance with

FIGS. 5B-5E depict cross sectional views of the MWA antenna system at cross section locations illustrated in FIG. 5A in accordance with an illustrative embodiment.

FIG. 6 depicts a top view of the MWA antenna system of accordance with an illustrative embodiment.

FIG. 7 shows a voltage standing wave ratio as a function of frequency of a simulated MWA antenna system illustrated in FIGS. 3 and 4 in accordance with an illustrative embodiment.

FIG. 8 shows a normalized specific absorption rate (SAR) pattern of the MWA antenna system of FIGS. 3 and 4 in the x-z plane in accordance with an illustrative embodiment.

FIG. 9 shows a normalized SAR pattern of the MWA antenna system of FIGS. 3 and 4 in the y-z plane in accordance with an illustrative embodiment.

FIG. 10 shows a 50° Celsius contour after 5 minutes of ablation using the MWA antenna system of FIGS. 3 and 4 with 20 Watts of input power in the x-z and y-z planes in accordance with an illustrative embodiment.

FIG. 11 depicts a top view of an MWA antenna system with a second balun in accordance with an illustrative embodiment.

FIG. 12 depicts a bottom view of the MWA antenna system of FIG. 11 in accordance with an illustrative embodiment.

FIG. 13 depicts a right side view of the MWA antenna system of FIG. 11 in accordance with an illustrative embodiment.

FIG. 14 depicts a left side view of the MWA antenna system of FIG. 11 in accordance with an illustrative embodiment.

FIG. **15**A depicts the top and right side views of the MWA antenna system of FIGS. **11** and **13** aligned in accordance with an illustrative embodiment.

FIGS. **15**B-**15**E depict cross sectional views of the MWA antenna system at cross section locations illustrated in FIG. **515**A in accordance with an illustrative embodiment.

FIG. 16 depicts a top view of the MWA antenna system of FIGS. 11-14 unrolled and placed on a flat surface in accordance with an illustrative embodiment.

FIG. **17** shows a simulated and a measured voltage ¹⁰ standing wave ratio as a function of frequency of the MWA antenna system illustrated in FIGS. **11-14** in accordance with an illustrative embodiment.

FIG. 18 shows a comparison between a simulated and a measured input reflection coefficient, S_{11} , of the MWA ¹⁵ antenna system of FIGS. 11-14 in accordance with an illustrative embodiment.

FIG. **19** shows a normalized SAR pattern of the MWA antenna system of FIGS. **11-14** in the x-z plane in accordance with an illustrative embodiment.

FIG. **20** shows a normalized SAR pattern of the MWA antenna system of FIGS. **11-14** in the y-z plane in accordance with an illustrative embodiment.

FIG. **21** shows a 50° Celsius contour after 5 minutes of ablation using the MWA antenna system of FIGS. **11-14** with ²⁵ 20 Watts of input power in the x-z and y-z planes in accordance with an illustrative embodiment.

FIG. 22 depicts a block diagram of an MWA system including the balun of FIGS. 3 and 4 or of FIGS. 11-14 in accordance with an illustrative embodiment. 30

DETAILED DESCRIPTION

With reference to FIG. 1, a microwave ablation (MWA) antenna 100 is connected to and fed by a coaxial cable 102 35 that provides electromagnetic energy to antenna 100 at a selected operating frequency f_0 . MWA can be used to provide thermal therapy for treatment of various types of cancer 108 in various tissue/organs 106. Tissue/organs 106 may include liver, kidney, lung, bone, etc. MWA uses 40 microwave frequency in the range 300 megahertz (MHz) to 300 gigahertz (GHz), though the range from 915 MHz to 2.45 GHz is most commonly used. MWA can be used to elevate the temperature of cancerous tissues to cytotoxic levels (e.g. >50° Celsius (C)) that quickly results in cell 45 death. Electromagnetic waves are introduced into cancerous tissues by inserting antenna 100 interstitially into the tumor or other cancerous tissue.

Typically, interstitial antennas used for MWA are implemented using coaxial cables. When a balanced antenna 50 (antenna 100) is fed by an unbalanced transmission line (coaxial cable 102), unwanted electric currents are excited on the outer conductor of coaxial cable 102. If not properly suppressed, the currents can result in undesired heating and ablation of healthy tissue along the insertion path of antenna 55 100. A balanced to unbalanced transformer (balun) may be implemented to suppress the currents.

With reference to FIG. 2, a widely used solution for choking the undesired outer surface currents for coaxially fed MWA antennas is to use a coaxial balun 200. Coaxial 60 balun 200 includes a cylindrical conductor with a circular cross section that encompasses coaxial cable 102. The cylindrical conductor and the outer conductor of coaxial cable 102 constitute a new coaxial transmission line for the outer surface currents. By properly choosing a length of 65 coaxial balun 200 and terminating its proximal end with a short circuit or an open circuit, a high impedance is pre-

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sented at its distal end to effectively choke the outer surface currents. However, this coaxial balun **200** increases an overall diameter, and therefore, invasiveness of antenna **100**.

With reference to FIG. 3, a top view of antenna 100 with a first balun 300 is shown in accordance with an illustrative embodiment. With reference to FIG. 4, a right side view of antenna 100 with first balun 300 is shown in accordance with an illustrative embodiment. Antenna 100 may be a dipole antenna formed by a first dipole arm 100a and a second dipole arm 306. Second dipole arm 306 may also be referred to herein as a prong. In the illustrative embodiment, second dipole arm 306 has a rectangular shape though other shapes may be used. Because second dipole arm 306 may be integrally formed with the remaining components of first balun 300, second dipole arm 306 is also referred to herein as a portion of first balun 300. First balun 300 is connected between coaxial cable 102 and first dipole arm 100a of antenna 100.

Antenna 100 is formed of a conductive material. As 20 understood by a person of skill in the art, the wavelength of operation, λ_0 , of antenna 100 is defined as $\lambda_0 = c/F_0$, where c is the speed of light in an environment in which antenna 100 is used, such as a body tissue, and f_0 is the selected operating frequency. For illustration, f_0 may be between 500 MHz and 25 30 GHz. A cross section of first dipole arm 100*a* may be circular, square, elliptical, rectangular, etc.

Coaxial cable 102 may include any length of coaxial cable having any characteristic impedance. Coaxial cable 102 may include a center conductor extending a length of coaxial cable 102, a dielectric material surrounding the center conductor along the length of coaxial cable 102, a conductive shield surrounding the dielectric material along the length of coaxial cable 102, and an insulating jacket surrounding the conductive shield along the length of coaxial cable 102 as understood by a person of skill in the art. The center conductor is generally circular and may be formed of a solid conductive material such as copper plated steel wire, silver plated steel wire, silver plated copper wire, silver plated copper clad steel wire, copper wire, copper clad aluminum wire, steel wire, etc. Coaxial cable 102 may have a variety of diameters. The dielectric material may include foamed polyethylene, solid polyethylene, polyethylene foam, polytetrafluoroethylene, air, air space polyethylene, vacuum, alumina, etc. For illustration, the dielectric material may include any low loss dielectric materials having a permittivity relative to a vacuum within the range of 1-10. The conductive shield may be formed of a solid or braided conductive material such as copper, steel, aluminum, silver plated copper, silver plated copper clad steel, etc. The insulating jacket (also known as an outer sheath) can be made from many different insulating materials such as polyvinyl chloride, polytetrafluoroethylene, another plastic material, etc.

Coaxial cable **102** may be rigid, semi-rigid, or flexible. The characteristic impedance may be off the shelf and range between approximately 20 and approximately 125 ohms (Ω) or may be designed to have a selected characteristic impedance within, above, or below this range as understood by a person of skill in the art using various dielectric and conductive materials, diameters, and thicknesses.

First balun 300 may include a tapered wall 302, a ring 304, and second dipole arm 306 that forms a prong connected to and extending from ring 304. First balun 300 is formed of a conductive material. For example, tapered wall 302, ring 304, and second dipole arm 306 may be formed of the same material as the conductive shield of coaxial cable 102. First balun 300 may be created by tapering the con-

ductive shield of a portion of coaxial cable **102** to a wall that connects to ring **304** and removing the material on each side of second dipole arm **306** except where it connects to ring **304** to form a slot and to make a smooth transition from a coaxial line to a parallel wire line.

Tapered wall **302** forms a portion of a cylindrical tube between a first wall **500** (shown referring to FIGS. **5**B and **6**) and a second wall **514** (shown referring to FIG. **5**E). First wall **500** is opposite second wall **514**. Tapered wall **302** surrounds a balun dielectric material **334**. Balun dielectric material **334** may surround a balun center conductor **326**. Balun center conductor **326** may be an extension of the center conductor of coaxial cable **102**. First dipole arm **100***a* may be an extension of balun center conductor **326**.

The portion of the cylindrical tube forms a slot exposing 15 balun dielectric material **334**. Balun dielectric material **334** may be the same material and/or an extension of the dielectric material that surrounds the center conductor of coaxial cable **102**. A second dielectric material **336** may further surround first dipole arm **100***a*. Second dielectric material 20 **336** may be the same material as and/or an extension of balun dielectric material **334**.

Tapered wall 302 has an exterior surface 308 and an interior surface 310. The slot formed by tapered wall 302 starts at a slot end point 312 and extends to a first end point 25 314 and a second end point 316 on either side of second wall 514. A first slot wall 318 extends from slot end point 312 to first end point 314. A second slot wall 320 extends from slot end point 312 to second end point 316.

Second dipole arm **306** extends from a first dipole arm 30 wall **332** that is closest to slot end point **312** to a second dipole arm wall **516** (shown referring to FIG. 5E) that mounts (connects) second dipole arm **306** to ring **304**. First dipole arm wall **332** is opposite second dipole arm wall **516**. A first arm wall **322** of second dipole arm **306** extends from 35 first dipole arm wall **332** to second dipole arm wall **516**. A second arm wall **332** to second dipole arm **306** extends from first dipole arm wall **332** to second dipole arm wall **516**. First arm wall **322** and second arm wall **324** are on opposite sides of second dipole arm **306** forms a 40 semicircular prong.

Ring 304 is connected to second wall 514 of tapered wall 302 and to second dipole arm wall 516 of second dipole arm 306. Ring 304 forms a short cylindrical tube surrounding balun center conductor 326 and balun dielectric material 45 334. Ring 304 has an inner ring wall 328 and an outer ring wall 330. Second dipole arm 306 is connected to inner ring wall 328 of ring 304, extends towards first wall 500, and is located in the slot created by tapered wall 302. Second dipole arm 100*a* extend in opposite 50 directions relative to ring 304. Second wall 514 of tapered wall 302 also connects to inner ring wall 328.

Tapered wall **302** and balun center conductor **326** form a balanced parallel-wire line to feed antenna **100** that includes first dipole arm **100***a* and second dipole arm **306**. Second 55 dipole arm **306** is placed in the slot formed by tapered wall **302** and is connected to tapered wall **302** through ring **304** at second dipole arm wall **516**. At its operating frequency, first balun **300** provides balanced currents for antenna **100**. As a result, unbalanced currents flowing back on an outer 60 surface of the conductive shield of coaxial cable **102** are minimized.

First balun 300 may be formed by removing a portion of the conductive shield of coaxial cable 102 to form the slot leaving tapered wall 302, second dipole arm 306, and ring 65 304. For example, a laser may be used to remove the portion of the conductive shield of coaxial cable 102. Thus, first

balun 300 may be formed from a portion of coaxial cable 102. As another example, first balun 300 may be etched from a hollow tube of conductive material such as copper and electrically connected to an inner surface of the conductive shield of coaxial cable 102.

A cover (insulating jacket or outer sheath) (not shown) may enclose first dipole arm 100a and first balun 300. The cover may be mounted to allow movement relative to first dipole arm 100a and first balun 300 so that first dipole arm 100a and first balun 300 are protected while inserted into a tissue, but can be exposed once inserted into the tissue.

With reference to FIG. 5A, the top and right side views of FIGS. 3 and 4 are aligned to show cross section locations of first balun 300 shown with reference to FIGS. 5B to 5E. With reference to FIG. 5B, a cross section of first balun 300 taken at slot end point 312 and indicated as 5B in FIG. 5A is shown in accordance with an illustrative embodiment. Balun center conductor 326 has a diameter 501. Balun dielectric material 334 has a dielectric width 502 that surrounds balun center conductor 326. First wall 500 of tapered wall 302 has a conductor width 504 that surrounds balun dielectric material 334 between exterior surface 308 and interior surface 310. A cross section of first balun 300 taken at outer ring wall 330 of ring 304 and also indicated as 5B in FIG. 5A is identical to that taken at slot end point 312 except that wall 500 of tapered wall 302 is replaced with outer ring wall 330 of ring 304.

With reference to FIG. **5**C, a cross section of first balun **300** taken at a point as indicated by **5**C in FIG. **5**A is shown in accordance with an illustrative embodiment. A first arc length **506** defines a slot arc length between first slot wall **318** and second slot wall **320** of tapered wall **302**.

With reference to FIG. 5D, a cross section of first balun 300 taken at first dipole arm wall 332 as indicated by 5D in FIG. 5A is shown in accordance with an illustrative embodiment. A second arc length 508 defines a slot arc length between first slot wall 318 and second slot wall 320 of tapered wall 302 for a cross section taken at first dipole arm wall 332. A prong arc length 510 defines an arc length between first arm wall 322 and second arm wall 324 of second dipole arm 306.

With reference to FIG. **5**E, a cross section of first balun **300** taken at second wall **514** of tapered wall **302** and at second dipole arm wall **516** as indicated by **5**E in FIG. **5**A is shown in accordance with an illustrative embodiment. A third arc length **512** defines a slot arc length between first slot wall **318** and second slot wall **320** at second wall **514** of tapered wall **302**.

With reference to FIG. 6, a top view of first balun 300 unrolled and placed on a flat surface is shown in accordance with an illustrative embodiment. A total length 600 of first balun 300 is measured between first wall 500 of tapered wall 302 and outer ring wall 330 of ring 304. A tapered wall length 602 of tapered wall 302 is measured between first wall 500 and second wall 514 of tapered wall 302. A tapered wall circumference 604 defines a circumference of tapered wall 302 at slot end point 312. A tapered wall height 606 is measured between first end point 314 and second end point 316 and defines a length of second wall 514 of tapered wall 302 that mounts to inner ring wall 328 of ring 304. A ring circumference 608 defines a circumference of inner ring wall 328 and outer ring wall 330 of ring 304. A ring width 610 defines a width of ring 304 measured between inner ring wall 328 and outer ring wall 330. A prong length 612 is measured between first dipole arm wall 332 and outer ring wall 330 of ring 304 and includes ring width 610. A prong width 614 defines a width of second dipole arm 306 mea-

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sured between first arm wall 322 and second arm wall 324. A dipole arm length 338 (shown referring to FIG. 3) is measured as a length of first dipole arm 100a.

In the illustrative embodiment, when flattened, first slot wall **318** and second slot wall **320** have a linear slope. In alternative embodiments, when flattened, first slot wall 318 and second slot wall 320 may have different slopes that may be non-linear. For example, when flattened, first slot wall 318 and second slot wall 320 may form a concave or convex curve between slot end point 312 and first end point and between slot end point 312 and second end point 316, respectively. First slot wall 318 and second slot wall 320 have a complementary shape.

Total length 600 may be selected from a range defined by 15

$$\frac{\lambda_0}{4}$$
 and $\frac{3\lambda_0}{2}$,

where λ_0 is the wavelength at the operating frequency of the signal carried by balun center conductor 326. Tapered wall height 606, ring width 610, and prong width 614 may each be selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

where d is diameter 501 of balun center conductor 326. Tapered wall height 606, ring width 610, and/or prong width 614 may be equal. Prong length 612 and dipole arm length 338 are selected from a range defined by

$$\frac{\lambda_1}{4}$$
 and $\frac{\lambda_1}{2}$,

where λ_1 is an effective wavelength of an operating fre-40 quency of a signal carried by balun center conductor 326 in a medium defined by a tissue into which first balun 300 and first dipole arm 100a are at least partially inserted. Tapered wall circumference 604 and ring circumference 608 may be equal to $2\pi r$, where

$$r = \frac{d}{2} + d_d + d_c,$$

where d_d equals dielectric width 502, and d_c equals conductor width 504.

For illustration, the parameters of first balun 300 and first dipole arm 100*a* to achieve localized specific absorption rate (SAR) and heating patterns and a good impedance match 55 between antenna 100 and coaxial cable 102 at 6 GHz were determined using full-wave electromagnetic (EM) simulations of the antennas and simplified thermal simulations. For example, the EM simulations of the antennas were conducted using CST Microwave Studio to design them to 60 operate at 6 GHz in liver tissue. Dielectric properties of liver at room temperature were modeled using a 1-pole Cole Cole model presented for the frequency range from 0 to 8 GHz. Absorption of electromagnetic fields in tissue, calculated from the EM simulations, was scaled for an input power of 65 20 W to be the heat source in transient thermal simulations in CST Multiphysics Suite. The dimensions were deter8

mined as prong length 612 equal to 6.2 millimeters (mm), dipole arm length 338 equal to 7 mm, total length 600 equal to 16 mm, tapered wall height 606, ring width 610, and prong width 614 each equal to 0.5 mm, and tapered wall circumference 604 and ring circumference 608 each equal to $\pi \times 2.2$ mm, the outer circumference of coaxial cable 102. Coaxial cable 102 was selected as $50-\Omega$ UT-085C semi-rigid cable. Copper tubing was used for first balun 300. Polytetrafluoroethylene was used for balun dielectric material 334.

Referring to FIG. 7, a voltage standing wave ratio (VSWR) curve 700 computed from the simulation results is shown. VSWR curve 700 shows a value less than two from four GHz to over eight GHz.

Referring to FIGS. 8 and 9, a normalized specific absorption rate (SAR) pattern of antenna 100 with first balun 300 is shown in the x-z and y-z planes, respectively, where z is along an axis of balun center conductor 326. The x-y planes define the cross sections of first balun 300 with the x-axis and the y-axis as indicated in FIG. 5D. Referring to FIG. 8, a -10 decibel (dB) curve 800 shows a SAR level reduced by 10 dB compared to a maximum SAR level. A -20 dB curve

802 shows a SAR level reduced by 20 dB compared to the maximum SAR level. A -30 dB curve 804 shows a SAR level reduced by 30 dB compared to the maximum SAR 25 level. A region 806 indicates a location of first balun 300 for reference.

Referring to FIG. 9, a -10 dB curve 900 shows a SAR level reduced by 10 dB compared to the maximum SAR level. A -20 dB curve 902 shows a SAR level reduced by 20 dB compared to the maximum SAR level. A -30 dB curve 904 shows a SAR level reduced by 30 dB compared to the maximum SAR level. A region 906 indicates a location of first balun 300 for reference. A proximal end 908 of second dipole arm 306 shows a reduction of approximately 20 dB 35 in SAR level. A proximal end 910 of first balun 300 shows a reduction of approximately 30 dB in SAR level. The results indicate that the outer surface currents are effectively suppressed using first balun 300, resulting in a compact SAR pattern with minimal tails along the shaft of coaxial cable 102.

An asymmetric heating pattern results due to a lack of symmetry in the placement of second dipole arm 306 that is connected to tapered wall 302 in the y-z plane. As a result, the electromagnetic radiation is stronger on the lower side, 45 where second dipole arm 306 is located, compared to an upper side where tapered wall 302 is located. Referring to FIG. 10, a simulated 50° C. contour, used to predict a boundary of an ablation zone, is shown in the x-z and the y-z planes at the end of a five minute ablation using an input power of 20 Watts (W). An x-z ablation zone 1000 is symmetrical in the x-z plane. A y-z ablation zone 1002 is slightly asymmetrical in the y-z plane, as a result of the corresponding SAR pattern in these two cut planes. However, maximum lateral dimensions of the ablation zone in the two cut planes are the same and equal to 24 mm. In the y-z plane, the maximum lateral radius of the ablation zone is 10.75 mm for the upper side (r_i) and 13.25 mm for the lower side (r_2) . While this degree of asymmetry is not significant, further modification of the antenna design to create a more asymmetric ablation zone can be implemented such as tilting first dipole arm 100a and second dipole arm 306 toward a lower half of the y-z plane or deploying a hemi-cylindrical metallic reflector to prevent EM radiation on the upper half of the y-z plane. Additionally, such asymmetrically enhanced heating may be desirable in certain clinical scenarios where tumors are highly asymmetric or where an antenna has to heat a tumor from a peripheral position

because the central region of the tumor is inaccessible (e.g. blockage by other vital organs).

With reference to FIG. 11, a top view of first dipole arm 100*a* with a second balun 1100 is shown in accordance with an illustrative embodiment. With reference to FIG. 12, a 5 bottom view of first dipole arm 100a with second balun 1100 is shown in accordance with an illustrative embodiment. With reference to FIG. 13, a right side view of first dipole arm 100a with second balun 1100 is shown in accordance with an illustrative embodiment. With reference to FIG. 14, 10 a left side view of first dipole arm 100a with second balun 1100 is shown in accordance with an illustrative embodiment. Antenna 100 may be a dipole antenna formed by first dipole arm 100a, a second dipole arm 1106a, and a third dipole arm 1106b. Second balun 1100 is similar to first balun 15 300 except that two slots are formed instead of one forming a symmetric structure with two slots opposite each other and with second dipole arm 1106a and third dipole arm 1106b opposite each other circumferentially around second balun 1100.

Second dipole arm 1106a and third dipole arm 1106b may each be referred to herein as a prong. Second dipole arm 1106a and third dipole arm 1106b are connected to a ring 1104 to form one arm of antenna 100 that again forms a dipole. Again, first dipole arm 100a is the second arm of the 25 dipole.

In the illustrative embodiment, second dipole arm 1106aand third dipole arm 1106b have a rectangular shape though other shapes may be used. Because second dipole arm 1106aand third dipole arm 1106b may be integrally formed with 30 the remaining components of second balun 1100, second dipole arm 1106a and third dipole arm 1106b are also referred to herein as portions of second balun 1100. Second balun 1100 is connected between coaxial cable 102 and first dipole arm 100a of antenna 100. 35

Second balun **1100** may include a tapered wall **1102**, ring **1104**, second dipole arm **1106***a*, and third dipole arm **1106***b* that form prongs connected to and extending from ring **1104**. Second balun **1100** is formed of a conductive material. For example, tapered wall **1102**, ring **1104**, second dipole arm **40 1106***a*, and third dipole arm **1106***b* may be formed of the same material as the conductive shield of coaxial cable **102**. Second balun **1100** may be created by tapering the conductive shield of a portion of coaxial cable **102** to form two walls that connect to ring **1104** and removing the material on 45 each side of second dipole arm **1106***a* and of third dipole arm **1106***b* except where each connects to ring **1104** to form two slots and to make a smooth transition from a coaxial line to a parallel wire line.

Tapered wall **1102** includes a first tapered wall **1108***a* and 50 a second tapered wall **1108***b*. First tapered wall **1108***a* and second tapered wall **1108***b* are similar to tapered wall **302**. First tapered wall **1108***a* and second tapered wall **1108***b* form two portions of a cylindrical tube between a first wall **1506** (shown referring to FIGS. **15B** and **16**) and a second 55 wall **1520***a* and a third wall **1520***b*, respectively (shown referring to FIG. **15E**). First wall **1506** is opposite second wall **1520***a* and third wall **1520***b*. First tapered wall **1108***a* and second tapered wall **1108***b* surround balun dielectric material **334**, which again surrounds balun center conductor 60 **326**. The portion of the cylindrical tube forms two slots exposing balun dielectric material **334**.

First tapered wall **1108***a* has an exterior surface **1508***a* (shown referring to FIG. **15**C) and an interior surface **1510***a* (shown referring to FIG. **15**C). First tapered wall **1108***a* 65 starts at a first slot end point **1118***a* on a first side and a second slot end point **1118***b* on a second side. A first slot wall

1120*a* of first tapered wall 1108*a* extends from second slot end point 1118*b* to first end point 1110*a*. A second slot wall 1122*a* of first tapered wall 1108*a* extends from first slot end point 1118*a* to second end point 1112*a*.

Second tapered wall **1108***b* has an exterior surface **1508***b* (shown referring to FIG. **15**C) and an interior surface **1510***b* (shown referring to FIG. **15**C). Second tapered wall **1108***b* starts at first slot end point **1118***a* on a first side and second slot end point **1118***b* on a second side. A third slot wall **1120***b* of second tapered wall **1108***b* extends from first slot end point **1118***a* to third end point **1110***b*. A fourth slot wall **1122***b* of second tapered wall **1108***b* extends from second slot end point **1118***b* to fourth end point **1112***b*.

Second dipole arm 1106a extends from a first dipole arm
15 wall 1115a that is closest to first slot end point 1118a to a second dipole arm wall 1522a (shown referring to FIG. 15E). First dipole arm wall 1115a is opposite second dipole arm wall 1522a. Second dipole arm wall 1522a mounts to ring 1104 between a first arm wall 1114a and a second arm
20 wall 1116a. Second dipole arm 1106a forms a semicircular prong.

Third dipole arm **1106***b* extends from a third dipole arm wall **1115***b* that is closest to second slot end point **1118***b* to a third dipole arm wall **1522***b* (shown referring to FIG. **15**E). Third dipole arm wall **1115***b* is opposite third dipole arm wall **1522***b*. Third dipole arm wall **1522***b* mounts to ring **1104** between a third arm wall **1114***b* and a fourth arm wall **1116***b*. Third dipole arm **1106***b* forms a semicircular prong.

Ring 1104 is similar to ring 304. Ring 1104 is connected to second wall 1520*a* of first tapered wall 1108*a*, to third wall 1520*b* of second tapered wall 1108*b*, to second dipole arm wall 1522*a* of second dipole arm 1106*a*, and to third dipole arm wall 1522*b* of third dipole arm 1106*b*. Ring 1104 forms a short cylindrical tube surrounding balun center conductor 326 and balun dielectric material 334. Ring 1104 has an inner ring wall 1124 and an outer ring wall 1126. Second dipole arm 1106*a* and third dipole arm 1106*b* are connected to inner ring wall 328 of ring 1104, extend towards first wall 1506, and are located in slots created by first tapered wall 1108*a* and second tapered wall 1108*b*, respectively. Second dipole arm 1106*a* and third dipole arm 1106*b* extend in a direction opposite to first dipole arm 100*a* relative to ring 1104.

First tapered wall **1108***a*, second tapered wall **1108***b*, and balun center conductor **326** form a balanced parallel-wire line to feed antenna **100** that includes first dipole arm **100***a*, second dipole arm **1106***a*, and third dipole arm **1106***b*. Second dipole arm **1106***a* and third dipole arm **1106***b* are placed in the slots formed by first tapered wall **1108***a* and second tapered wall **1108***b* and are connected to tapered wall **302** through ring **1104** at second dipole arm wall **1522***a* and at third dipole arm wall **1522***b*, respectively. At its operating frequency, second balun **1100** provides balanced currents for antenna **100**. As a result, unbalanced currents flowing back on an outer surface of the conductive shield of coaxial cable **102** are minimized.

Second balun **1100** may be formed by removing a portion of the conductive shield of coaxial cable **102** to form the two slots, second dipole arm **1106**a, and third dipole arm **1106**b. For example, a laser may be used to remove the portion of the conductive shield of coaxial cable **102**. As another example, second balun **1100** may be etched from a hollow tube of conductive material such as copper and electrically connected to an inner surface of the conductive shield of coaxial cable **102**.

A cover (not shown) may enclose first dipole arm **100***a* and second balun **1100**. The cover may be mounted to allow

movement relative to first dipole arm 100a and second balun 1100 so that first dipole arm 100a and second balun 1100 are protected while inserted into a tissue, but can be exposed once inserted into the tissue.

With reference to FIG. 15A, the top and right side views 5 of FIGS. 11 and 13 are aligned to show cross section locations of second balun 1100 shown with reference to FIGS. 15B to 15E. With reference to FIG. 15B, a cross section of second balun 1100 taken at first slot end point 1118*a* and at second slot end point 1118*b* and indicated as 10 15B in FIG. 15A is shown in accordance with an illustrative embodiment. Balun center conductor 326 has a diameter 1500. Balun dielectric material 334 has a dielectric width 1502 that surrounds balun center conductor 326. Tapered wall 1102 has a conductor width 1504 that surrounds balun 15 dielectric material 334. A cross section of second balun 1100 taken at outer ring wall 1126 of ring 1104 is identical to that taken at first slot end point 1118a and at second slot end point 1118b. A cross section of second balun 1100 taken at outer ring wall 1126 of ring 1104 is identical to that taken at 20 first slot end point 1118a except that first wall 1506 of tapered wall 1102 is replaced with outer ring wall 1126 of ring 1104.

With reference to FIG. 15C, a cross section of second balun 1100 taken at a point as indicated by 15C in FIG. 15A 25 is shown in accordance with an illustrative embodiment. A first arc length 1512 defines an arc length between first slot wall 1120*a* and second slot wall 1122*a* of first tapered wall 1108*a*. An arc length (not shown) for second tapered wall 1108*b* is identical to first arc length 1512. Conductor width 30 1504 is identical between a first exterior surface 1508*a* and a first interior surface 1510*a* of first tapered wall 1108*b* and between a second exterior surface 1508*b* and a second interior surface 1510*b* of second tapered wall 1108*b*.

With reference to FIG. 15D, a cross section of second 35 balun 1100 taken at first dipole arm wall 1115*a* and at second dipole arm wall 1115*b* as indicated by 15D in FIG. 15A is shown in accordance with an illustrative embodiment. A second arc length 1514 defines an arc length between first slot wall 1120*a* and second slot wall 1122*a* of first tapered 40 wall 1108*a* for a cross section taken at first dipole arm wall 1115*a*. An arc length (not shown) for second tapered wall 1108*b* is identical to second arc length 1514 at first dipole arm wall 1115*a*. A prong arc length 1516 defines an arc length between first arm wall 1114*a* and second arm wall 45 1116*a* of second dipole arm 1106*a* and between third arm wall 1114*b* and fourth arm wall 1116*b* of third dipole arm 1106*b*.

With reference to FIG. 15E, a cross section of second balun 1100 taken at second wall 1520*a*, third wall 1520*b*, 50 second dipole arm wall 1522*a*, and third dipole arm wall 1522*b* as indicated by 15E in FIG. 15A is shown in accordance with an illustrative embodiment. A minimum arc length 1518 defines an arc length between first slot wall 1120*a* and second slot wall 1122*a* of first tapered wall 55 1108*a*. An arc length (not shown) for second tapered wall 1108*b* is identical to minimum arc length 1518 at first dipole arm wall 1115*a*.

With reference to FIG. 16, a top view of second balun 1100 unrolled and placed on a flat surface is shown in 60 accordance with an illustrative embodiment. A total length 1600 of second balun 1100 is measured between first wall 1506 and outer ring wall 1126 of ring 1104. A tapered wall length 1602 of first tapered wall 1108*a* and of second tapered wall 1108*b* is measured between first wall 1506 and second 65 wall 1520*a* of first tapered wall 1108*a* and between first wall 1506 and fourth wall 1520*b* of second tapered wall 1108*b*, 12

respectively. A tapered wall circumference 1604 defines a circumference of tapered wall 1102 at first slot end point 1115a. A tapered wall height 1606 is measured between first end point 1110a and second end point 1112a and defines a length of second wall 1520a of first tapered wall 1108a that mounts to inner ring wall 1124 of ring 1104. Tapered wall height 1606 also defines a length of third wall 1520b of second tapered wall 1108b that mounts to inner ring wall 1124 of ring 1104. A ring circumference 1608 defines a circumference of inner ring wall 1124 and outer ring wall 1126 of ring 1104. A ring width 1610 defines a width of ring 1104 measured between inner ring wall 1124 and outer ring wall 1126. A prong length 1612 is measured between first dipole arm wall 1115a and outer ring wall 330 of ring 1104 and includes ring width 1610. Prong length 1612 also defined a length between second dipole arm wall 1115b and outer ring wall 330 of ring 1104 and includes ring width 1610. A prong width 1614 defines a width of second dipole arm 1106a measured between first arm wall 1114a and second arm wall 1116a. A prong width 1614 defines a width of third dipole arm 1106b measured between third arm wall 1114b and fourth arm wall 1116b.

In the illustrative embodiment, when flattened, first slot wall **1120***a*, second slot wall **1122***a*, third slot wall **1120***b*, and fourth slot wall **1122***b* have a linear slope. In alternative embodiments, when flattened, first slot wall **1120***a*, second slot wall **1122***a*, third slot wall **1120***b*, and fourth slot wall **1122***a*, third slot wall **1120***b*, and fourth slot wall **1122***b* may have different slopes that may be non-linear. For example, when flattened, first slot wall **1120***a*, second slot wall **1122***a*, third slot wall **1120***b*, and fourth slot wall **1122***b* may form a concave or convex curve between the slot wall end points. First slot wall **1120***a* and third slot wall **1120***b* have a same flattened shape. Second slot wall **1122***a* and fourth slot wall **1122***b* have a same flattened shape that is complementary to that of first slot wall **1120***a* and third slot wall **1120***b*.

Total length 1600 may be selected from a range defined by

$$\frac{\lambda_0}{4}$$
 and $\frac{3\lambda_0}{2}$,

where λ_0 is the wavelength at the operating frequency of the signal carried by balun center conductor **326**. Tapered wall height **1606**, ring width **1610**, and prong width **1614** may each be selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

where d is diameter **1500** of balun center conductor **326**. Tapered wall height **1606**, ring width **1610**, and/or prong width **1614** may be equal. Prong length **1612** and dipole arm length **338** are selected from a range defined by

$$\frac{\lambda_1}{4}$$
 and $\frac{\lambda_1}{2}$,

where λ_1 is an effective wavelength of an operating frequency of a signal carried by balun center conductor **326** in a medium defined by a tissue into which second balun **1100** and first dipole arm **100***a* are at least partially inserted. Tapered wall circumference **1604** and ring circumference **1608** may be equal to $2\pi r$, where

$$r = \frac{d}{2} + d_d + d_c,$$

where d_d equals dielectric width 1502, and d_c equals conductor width 1504.

For illustration, the parameters of second balun 1100 and first dipole arm 100a to achieve localized specific absorption rate (SAR) and heating patterns and a good impedance match between antenna 100 and coaxial cable 102 at 6 GHz were again determined. The dimensions were determined as prong length 1612 equal to 8 mm, dipole arm length 1128 equal to 7 mm, total length 1600 equal to 18 mm, prong width 1614 equal to 0.7 mm, tapered wall height 1606 and 15 ring width 1610 each equal to 0.5 mm, and tapered wall circumference 1604 and ring circumference 1608 each equal to $\pi \times 2.2$ mm, the outer circumference of coaxial cable 102. Coaxial cable 102 was selected as $50-\Omega$ UT-085C semi-rigid cable. Copper tubing was used for second balun 1100. 20 Polytetrafluoroethylene was used for balun dielectric material 334. Second balun 1100 and first dipole arm 100a were embedded in a polytetrafluoroethylene coating with a diameter of 2.6 mm.

Referring to FIG. 17, a measured VSWR curve 1700 and 25 a simulated VSWR curve 1702 are shown. Referring to FIG. 18, a measured $|S_{11}|$ curve 1800 and a simulated $|S_{11}|$ curve 1802 are shown. The measured and simulated curves are in good agreement and show a good impedance match between antenna 100 and coaxial cable 102 over a wide frequency 30 range (from 5 GHz to over 8 GHz). At the operating frequency of 6 GHz, the measured VSWR is 1.19 ($|S_{11}|$ of -22 dB), which is slightly better than the simulated values (VSWR=1.21, $|S_{11}|=-20$ dB). Additionally, the measured input impedance of antenna 100 was unchanged as the 35 insertion depth of antenna 100 was varied as long as the entire second balun 1100 was immersed in liver with first dipole arm 100a. This confirmed that the outer surface currents were effectively suppressed along coaxial cable 102 up to a starting point of second balun 1100.

First dipole arm **100***a* with second balun **1100** achieved a slightly better impedance matching (VSWR=1.21, $|S_{11}|=-20 \text{ dB}$) than first dipole arm **100***a* with first balun **300** (VSWR=1.38, $|S_{11}|=-16 \text{ dB}$) at the operating frequency of 6 GHz. While increasing tapered wall length **1602** of first 45 tapered wall **1108***a* and of second tapered wall **1108***b* may help reduce the reflection coefficient, the frequency of best impedance match is most sensitive to the dimensions of first dipole arm **100***a*, second dipole arm **1106***a*, and third dipole arm **1106***b* of the dipole and a thickness of the outermost 50 coating, which in this case was 2.6 mm of polytetrafluoro-ethylene.

Referring to FIGS. **19** and **20**, a normalized SAR pattern of antenna **100** with second balun **1100** is shown in the x-z and y-z planes, respectively, where z is along an axis of 55 balun center conductor **326**. The x-y planes define the cross sections of second balun **1100** with the x-axis and the y-axis as indicated in FIG. **5**D. Referring to FIG. **19**, a -10 dB curve **1900** shows a SAR level reduced by 10 dB compared to the maximum SAR level. A -20 dB curve **1902** shows a 60 SAR level reduced by 20 dB compared to the maximum SAR level. A -30 dB curve **1904** shows a SAR level reduced by 30 dB compared to the maximum SAR level. A region **1906** indicates a location of second balun **1100** for reference.

Referring to FIG. **20**, a –10 dB curve **2000** shows a SAR 65 level reduced by 10 dB compared to the maximum SAR level. A –20 dB curve **2002** shows a SAR level reduced by

20 dB compared to the maximum SAR level. A -30 dB curve 2004 shows a SAR level reduced by 30 dB compared to the maximum SAR level. A region 2006 indicates a location of second balun 1100 for reference. The results indicate that the outer surface currents are effectively suppressed using second balun 1100, resulting in a compact SAR pattern with minimal tails along the shaft of coaxial cable 102. SAR values are reduced by 30 dB near first slot end point 1115*a* and second slot end point 1115*b*. First dipole arm 100*a* with second balun 1100 produced a highly compact volume enclosed by -30 dB curve 2004, of which the lateral diameter (about 26 mm) is even larger than the axial diameter (about 20 mm). Due to the lack of an axial symmetry, the SAR pattern in the x-z plane is slightly different from that in the y-z plane.

Referring to FIG. **21**, a simulated 50° C. contour, used to predict a boundary of an ablation zone, is shown in the x-z and the y-z planes at the end of a five minute ablation using an input power of 20 Watts (W). An x-z ablation zone **2100** and a y-z ablation zone **2102** are symmetrical. Despite a slight difference of the SAR pattern in the x-z (**2100**) and y-z (**2102**) planes, the ablation zone appears to be identical in these two cut planes, which demonstrates that first dipole arm **100***a* with second balun **1100** is capable of producing a rotationally symmetric ablation zone.

Compared to the antenna using the single-slot balun, SAR values less than -30 dB fall off faster along the shaft of second balun 1100. This is evident in the slightly longer tails of the -30 dB contours shown in -30 dB curve 804 and -30 dB curve 904 in comparison with -30 dB curve 1904 and -30 dB curve 2004. Moreover, first dipole arm 100a with second balun 1100 produced a symmetric SAR pattern in the y-z plane, as opposed to the asymmetric one in this cut plane of first dipole arm 100a with first balun 300. While thermal simulation results show that both provide ablation zones with similar dimensions, the one provided by first dipole arm 100a with second balun 1100 is more rotationally symmetric. Overall, the better impedance matching and more sym-40 metric heating pattern make first dipole arm 100a with second balun 1100 a more desirable design for ablation applications where directional heating is not needed.

Referring to FIG. 22, a block diagram of a MWA system 2200 is shown in accordance with an illustrative embodiment. MWA system 2200 may include a signal generator 2202, an amplifier 2204, a connector 2206, coaxial cable 102, first balun 300 or second balun 1100, and antenna 100. For illustration, antenna 100 may include first dipole arm 100a and second dipole arm 1106a and third dipole arm 1106b. Signal generator 2202 generates an analog signal at the operating frequency selected for antenna 100. A duty cycle of the analog signal may be controlled by signal generator 2202 based, for example, on an ablation zone size and heating rate. The analog signal may be amplified by amplifier 2204. Connector 2206 connects a second end of coaxial cable 102 opposite first dipole arm 100a. The loss through coaxial cable 102 is considered when adjusting the output power level of amplifier 2204 for a desired input power level to first dipole arm 100a and second dipole arm 1106a and third dipole arm 1106b. Connector 2206 may be a coaxial connector designed to maintain the coaxial form across the connection and having the same impedance as coaxial cable 102.

Use of directional terms, such as top, bottom, right, left, front, back, upper, lower, horizontal, vertical, behind, etc. are merely intended to facilitate reference to the various surfaces of the described structures relative to the orientations introduced in the drawings and are not intended to be limiting in any manner unless otherwise indicated.

As used in this disclosure, the term "connect" includes join, unite, mount, couple, associate, insert, hang, hold, affix, attach, fasten, bind, paste, secure, bolt, screw, rivet, pin, nail, 5 clasp, clamp, cement, fuse, solder, weld, glue, form over, slide together, layer, and other like terms. The phrases "connected on" and "connected to" include any interior or exterior portion of the element referenced. Elements referenced as connected to each other herein may further be 10 integrally formed together. As a result, elements described herein as being connected to each other need not be discrete structural elements. The elements may be connected permanently, removably, or releasably.

As used in this disclosure, the term "mount" includes join, 15 unite, connect, couple, associate, insert, hang, hold, affix, attach, fasten, bind, paste, secure, bolt, screw, rivet, pin, nail, clasp, clamp, cement, fuse, solder, weld, glue, form over, slide together, layer, and other like terms. The phrases "mounted on" and "mounted to" include any interior or 20 wall is selected from a range defined by exterior portion of the element referenced. These phrases also encompass direct connection (in which the referenced elements are in direct contact) and indirect connection (in which the referenced elements are not in direct contact, but are mounted together via intermediate elements). Elements 25 referenced as mounted to each other herein may further be integrally formed together. As a result, elements described herein as being mounted to each other need not be discrete structural elements. The elements may be mounted permanently, removably, or releasably.

The word "illustrative" is used herein to mean serving as an example, instance, or illustration. Any aspect or design described herein as "illustrative" is not necessarily to be construed as preferred or advantageous over other aspects or designs. Further, for the purposes of this disclosure and 35 unless otherwise specified, "a" or "an" means "one or more". Still further, using "and" or "or" in the detailed description is intended to include "and/or" unless specifically indicated otherwise.

The foregoing description of illustrative embodiments of 40 the disclosed subject matter has been presented for purposes of illustration and of description. It is not intended to be exhaustive or to limit the disclosed subject matter to the precise form disclosed, and modifications and variations are possible in light of the above teachings or may be acquired 45 from practice of the disclosed subject matter. The embodiments were chosen and described in order to explain the principles of the disclosed subject matter and as practical applications of the disclosed subject matter to enable one skilled in the art to utilize the disclosed subject matter in 50 various embodiments and with various modifications as suited to the particular use contemplated. It is intended that the scope of the disclosed subject matter be defined by the claims appended hereto and their equivalents.

What is claimed is:

- 1. A balun comprising:
- a center conductor extending a length of the balun;
- a dielectric material surrounding the center conductor along the length of the balun;
- a tapered wall forming a portion of a tube between a first wall and a second wall, wherein the first wall is opposite the second wall, wherein the tapered wall is formed of a conductive material, wherein the portion of the tube forms a slot exposing the dielectric material; 65
- a ring connected to the second wall of the tapered wall, wherein the ring is formed of the conductive material,

wherein the ring forms a tube surrounding the center conductor and the dielectric material; and

a prong connected to the ring to extend toward the first wall, wherein the prong is formed of the conductive material, wherein the prong extends over a portion of the dielectric material exposed by the slot.

2. The balun of claim 1, wherein a length between the first wall and an outer wall of the ring is selected from a range defined by

$$\frac{\lambda_1}{4}$$
 and $\frac{3\lambda_1}{2}$,

where λ_1 is a wavelength at an operating frequency of a signal carried by the center conductor, wherein the outer wall is opposite the second wall.

3. The balun of claim 2, wherein a width of the second

$$\frac{d}{2}$$
 and $2 \times d$,

where d is a diameter of the center conductor, wherein the width of the second wall is in a direction perpendicular to the length.

4. The balun of claim 1, wherein a length between a wall of the prong closest to the first wall and an outer wall of the ring is selected from a range defined by $\lambda_{1/4}$ and $\lambda_{1/2}$, where λ_1 is a wavelength at an operating frequency of a signal carried by the center conductor, wherein the outer wall is opposite the second wall.

5. The balun of claim 4, wherein a width of the prong is selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

where d is a diameter of the center conductor, wherein the width of the prong is in a direction perpendicular to the length.

6. The balun of claim 1, wherein a length between the second wall and an outer wall of the ring is selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

60

55 where d is a diameter of the center conductor, wherein the outer wall is opposite the second wall.

7. The balun of claim 1, wherein a circumference of the ring is equal to a length of the first wall.

8. The balun of claim 1, further comprising:

a second tapered wall forming a second portion of the tube between a third wall and a fourth wall, wherein the second tapered wall is formed of the conductive material, wherein the second portion of the tube forms a second slot exposing the dielectric material, wherein the ring is connected to the fourth wall of the second tapered wall, wherein the third wall is opposite the fourth wall; and

5

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a second prong connected to the ring to extend toward the third wall, wherein the second prong is formed of the conductive material, wherein the second prong extends over a second portion of the dielectric material exposed by the second slot.

9. The balun of claim 8, wherein a length between the first wall and an outer wall of the ring is selected from a range defined by

$$\frac{\lambda_1}{4}$$
 and $\frac{3\lambda_1}{2}$,

where λ_1 is a wavelength at an operating frequency of a 15 signal carried by the center conductor, wherein the outer wall is opposite the second wall, wherein a second length between the third wall and the outer wall of the ring is equal to the length.

10. The balun of claim **9**, wherein a width of the second 20 wall is selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$, 25

where d is a diameter of the center conductor, wherein the width of the second wall is in a direction perpendicular to the length, wherein a width of the fourth wall is equal to the width of the second wall. 30

11. The balun of claim 8, wherein a length between a wall of the prong closest to the first wall and an outer wall of the ring is selected from a range defined by $\lambda_{1}/4$ and $\lambda_{1}/2$, where λ_{1} is a wavelength at an operating frequency of a signal carried by the center conductor, wherein the outer wall is ³⁵ opposite the second wall, wherein a second length between a wall of the second prong closest to the third wall and the outer wall of the ring is equal to the length.

12. The balun of claim **11**, wherein a width of the prong is selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

where d is a diameter of the center conductor, wherein the width of the prong is in a direction perpendicular to the length, wherein a width of the second prong is equal to the width of the prong.

13. The balun of claim **8**, wherein a length between the second wall and an outer wall of the ring is selected from a range defined by

$$\frac{d}{2}$$
 and $2 \times d$,

where d is a diameter of the center conductor, wherein the outer wall is opposite the second wall.

14. The balun of claim 8, wherein a circumference of the ring is equal to a length of the first wall plus the third wall.

15. An antenna system comprising:

- a coaxial cable comprising
- a center conductor extending a length of the coaxial cable;
- a dielectric material surrounding the center conductor along the length of the coaxial cable; and
- a conductive shield surrounding the dielectric material along the length of the coaxial cable;
- a first dipole arm connected to the center conductor; and a balun connected between an end of the coaxial cable and
- the first dipole arm, the balun comprising a second center conductor extending a length of the
- a second center conductor extending a length of the balun, the second center conductor connected to and extending from the center conductor between the center conductor and the first dipole arm;
- a second dielectric material surrounding the second center conductor along the length of the balun;
- a tapered wall forming a portion of a tube between a first wall and a second wall, wherein the first wall is opposite the second wall, wherein the first wall is connected to the conductive shield, wherein the tapered wall is formed of a conductive material, wherein the portion of the tube forms a slot exposing the second dielectric material;
- a ring connected to the second wall of the tapered wall, wherein the ring is formed of the conductive material, wherein the ring forms a tube surrounding the second center conductor and the second dielectric material; and
- a prong connected to the ring to extend toward the first wall, wherein the prong is formed of the conductive material, wherein the prong extends over a portion of the second dielectric material exposed by the slot and forms a second dipole arm.

16. The antenna system of claim 15, wherein the second center conductor is formed as an extension of the center conductor.

17. The antenna system of claim 16, wherein the first dipole arm is formed as an extension of the second center conductor.

18. The antenna system of claim **15**, wherein a length between the first wall and an outer wall of the ring is selected from a range defined by

$$\frac{\lambda_1}{4}$$
 and $\frac{3\lambda_1}{2}$,

45

50

60

where λ_1 is a wavelength at an operating frequency of a signal carried by the center conductor, wherein the outer wall is opposite the second wall.

19. The antenna system of claim 15, wherein a length between a wall of the prong closest to the first wall and an outer wall of the ring is selected from a range defined by $\lambda_{1/4}$ and $\lambda_{1/2}$, where λ_{1} is a wavelength at an operating frequency of a signal carried by the center conductor, wherein the outer wall is opposite the second wall.

20. The antenna system of claim **15**, wherein a length of the first dipole arm is selected from a range defined by $\lambda_1/4$ and $\lambda_1/2$, where λ_1 is a wavelength at an operating frequency of a signal carried by the center conductor.

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